Research Article

Master-Slave Synchronization of Stochastic Neural Networks with Mixed Time-Varying Delays

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This paper investigates the problem on master-salve synchronization for stochastic neural networks with both time-varying and distributed time-varying delays. Together with the drive-response concept, LMI approach, and generalized convex combination, one novel synchronization criterion is obtained in terms of LMIs and the condition heavily depends on the upper and lower bounds of state delay and distributed one. Moreover, the addressed systems can include some famous network models as its special cases, which means that our methods extend those present ones. Finally, two numerical examples are given to demonstrate the effectiveness of the presented scheme.

1. Introduction

In the past decade, synchronization of chaotic systems has attracted considerable attention since the pioneering works of Pecora and Carroll [1], in which it shows when some conditions are satisfied, a chaotic system (the slave/response system) may become synchronized to another identical chaotic system (the master/drive system) if the master system sends some driving signals to the slave one. Now, it is widely known that there exist many benefits of having synchronization or chaos synchronization in various engineering fields, such as secure communication [2], image processing [3], and harmonic oscillation generation. Meanwhile, there exists synchronization in language development, which comes up with a common vocabulary, while agents' synchronization in organization management will improve their work efficiency. Recently, chaos synchronization has been widely investigated due to its

great potential applications. Especially, since artificial neural network model can exhibit the chaotic behaviors [4, 5], the synchronization has become an important area of study, see [6–23] and references therein. As special complex networks, delayed neural networks have been also found to exhibit some complex and unpredictable behaviors including stable equilibria, periodic oscillations, bifurcation, and chaotic attractors [24–27]. Presently, many literatures dealing with chaos synchronization phenomena in delayed neural networks have appeared. Together with various techniques such as LMI tool, *M*-matrix, and Jensen's inequalities, some elegant results have been derived for global synchronization of various delayed neural networks including discrete-time ones in [6–14]. Moreover, some authors have considered the problems on adaptive synchronization and H_{∞} synchronization in [15, 16].

Meanwhile, it is worth noting that, like time-delay and parameter uncertainties, noises are ubiquitous in both nature and man-made systems and the stochastic effects on neural networks have drawn much particular attention. Thus a large number of elegant results concerning dynamics of stochastic neural networks have already been presented in [17–23, 28, 29]. Since noise can induce stability and instability oscillations to the system, by virtue of the stability theory for stochastic differential equations, there has been an increasing interest in the study of synchronization for delayed neural networks with stochastic perturbations [17–23]. Based on LMI technique, in [17–19], some novel results have been derived on the global synchronization as the addressed networks were involved in distributed delay or neutral type. Also the works [20-23] have considered the adaptive synchronization and lag synchronization for stochastic delayed neural networks. However, the control schemes in [17–19] cannot tackle the cases as the upper bound of delay's derivative is not less than 1, and the presented results in [20-23] are not formulated in terms of LMIs, which makes them checked inconveniently by most recently developed algorithms. Meanwhile, in order to implement the practical point of view better, distributed delay should be taken into consideration and thus, some researchers have began to give some preliminary discussions in [9–11, 19]. It is worth pointing out that the range of time delays considered in [17– 23] is from 0 to an upper bound. In practice, the range of delay may vary in a range for which the lower bound is not restricted to be 0. Thus the criteria in the above literature can be more conservative because they have not considered the information on the lower bound of delay. Meanwhile, it has been verified that the convex combination idea was more efficient than some previous techniques when tackling time-varying delay, and furthermore, the novel idea needs some improvements since it has not taken distributed delay into consideration altogether [30]. Yet, few authors have employed improved convex combination to consider the stochastic neural networks with both variable and distributed variable delays and proposed some less conservative and easy-to-test control scheme for the exponential synchronization, which constitutes the main focus of the presented work.

Motivated by the above-mentioned discussion, this paper focuses on the exponential synchronization for a broad class of stochastic neural networks with mixed time-varying delays, in which two involved delays belong to the intervals. The form of addressed networks can include several well-known neural network models as the special cases. Together with the drive-response concept and Lyapunov stability theorem, a memory control law is proposed which guarantees the exponential synchronization of the drive system and response one. Finally, two illustrative examples are given to illustrate that the obtained results can improve some earlier reported works.

Notation 1. For symmetric matrix X, X > 0 (resp., $X \ge 0$) means that X > 0 ($X \ge 0$) is a positive-definite (resp., positive-semidefinite) matrix; A^T, A^{-T} represent the transposes of

matrices *A* and *A*⁻¹, respectively. For $\tau > 0$, $C([-\tau, 0]; \mathbf{R}^n)$ denotes the family of continuous functions φ from $[-\tau, 0]$ to \mathbf{R}^n with the norm $\|\varphi\| = \sup_{\tau \le \theta \le 0} |\varphi|$. Let $(\Omega, \mathcal{F}, \{\mathcal{F}_t\}_{t \ge 0}, P)$ be a complete probability space with a filtration $\{\mathcal{F}_t\}_{t \ge 0}$ satisfying the usual conditions; $L^p_{\mathcal{F}_0}([-\tau, 0]; \mathbf{R}^n)$ is the family of all \mathcal{F}_0 -measurable $C([-\tau, 0]; \mathbf{R}^n)$ -valued random variables $\xi = \{\xi(\theta) : -\tau \le \theta \le 0\}$ such that $\sup_{-\tau \le \theta \le 0} \mathbf{E} |\xi(\theta)|^p < \infty$, where $\mathbf{E}\{\cdot\}$ stands for the mathematical expectation operator with respect to the given probability measure *P*; *I* denotes the identity matrix with an appropriate dimension and $\begin{bmatrix} X & Y \\ Y^T & Z \end{bmatrix} = \begin{bmatrix} X & Y \\ X & Z \end{bmatrix}$ with * denoting the symmetric term in a symmetric matrix.

2. Problem Formulations

Consider the following stochastic neural networks with time-varying delays described by

$$dz(t) = \left[-b(z(t)) + Ag(z(t)) + Bg(z(t - \tau(t))) + D \int_{t-\varrho(t)}^{t} g(z(s))ds + I \right] dt, \qquad (2.1)$$

where $z(t) = [z_1(t), ..., z_n(t)]^T \in \mathbf{R}^n$ is the neuron state vector, $g(z(\cdot)) = [g_1(z_1(\cdot)), ..., g_n(z_n(\cdot))]^T \in \mathbf{R}$ represents the neuron activation function, $\mathbf{I} \in \mathbf{R}^n$ is a constant external input vector, and A, B, D are the connection weight matrix, the delayed weight matrix, and the distributively delayed connection weight one, respectively.

In the paper, we consider the system (2.1) as the master system and the slave system as follows:

$$dy(t) = \left[-b(y(t)) + Ag(y(t)) + Bg(y(t - \tau(t))) + D \int_{t-\varrho(t)}^{t} g(y(s))ds + I + u(t) \right] dt$$

+ $\sigma(t, \varepsilon(t), \varepsilon(t - \tau(t)))dw(t)$ (2.2)

with $\varepsilon(t) = [\varepsilon_1(t), \dots, \varepsilon_n(t)]^T = y(t) - z(t)$, where *A*, *B*, *D* are constant matrices similar to the relevant ones (2.1) and u(t) is the appropriate control input that will be designed in order to obtain a certain control objective. In practical situations, the output signals of the drive system (2.1) can be received by the response one (2.2).

The following assumptions are imposed on systems (2.1) and (2.2) throughout the paper.

(A1) Here $\tau(t)$ and $\varrho(t)$ denote the time-varying delay and the distributed one satisfying

$$0 \le \tau_0 \le \tau(t) \le \tau_m, \qquad \dot{\tau}(t) \le \mu, \ 0 \le \varrho_0 \le \varrho(t) \le \varrho_m, \tag{2.3}$$

and we introduce $\overline{\tau}_m = \tau_m - \tau_0$, $\overline{\varrho}_m = \varrho_m - \varrho_0$, and $\tau_{\max} = \max{\{\tau_m, \varrho_m\}}$.

(A2) Each function $b_i(\cdot) : \mathbf{R} \to \mathbf{R}$ is locally Lipschitz, and there exist positive scalars π_i and γ_i (i = 1, 2, ..., n) such that $\pi_i \ge \dot{b}_i(z) \ge \gamma_i > 0$ for all $z \in \mathbf{R}$. Here, we denote $\Pi = \text{diag}\{\pi_1, ..., \pi_n\}$ and $\Gamma = \text{diag}\{\gamma_1, ..., \gamma_n\}$. (A3) For the constants σ_i^+, σ_i^- , the neuron activation functions in (2.1) are bounded and satisfy

$$\sigma_i^- \le \frac{g_i(x) - g_i(y)}{x - y} \le \sigma_i^+, \quad \forall x, y \in \mathbf{R}, \ x \ne y, \ i = 1, 2, \dots, n.$$
(2.4)

(A4) In system (2.2), the function $\sigma(t, \cdot, \cdot) : \mathbf{R}^+ \times \mathbf{R}^n \times \mathbf{R}^n \to \mathbf{R}^{n \times m}$ ($\sigma(t, 0, 0) = 0$) is locally Lipschitz continuous and satisfies the linear growth condition as well. Moreover, $\sigma(t, \cdot, \cdot)$ satisfies the following condition:

trace
$$\left[\sigma^{T}(t,x,y)\sigma(t,x,y)\right] \leq x^{T}\Pi_{1}^{T}\Pi_{1}x + y^{T}\Pi_{2}^{T}\Pi_{2}y, \quad \forall x,y \in \mathbf{R}^{n},$$
 (2.5)

where Π_i (*i* = 1, 2) are the known constant matrices of appropriate dimensions.

Let $\varepsilon(t)$ be the error state and subtract (2.1) from (2.2); it yields the synchronization error dynamical systems as follows:

$$d\varepsilon(t) = \left[-\beta(\varepsilon(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t))) + D\int_{t-\varrho(t)}^{t} f(\varepsilon(s))ds + u(t)\right]dt + \sigma(t,\varepsilon(t),\varepsilon(t - \tau(t)))dw(t),$$
(2.6)

where $f(\varepsilon(\cdot)) = g(y(\cdot)) - g(z(\cdot))$. One can check that the function $f_i(\cdot)$ satisfies $f_i(0) = 0$, and

$$\sigma_i^- \le \frac{f_i(x) - f_i(y)}{x - y} \le \sigma_i^+, \quad \forall x, y \in \mathbf{R}, \ x \ne y, \ i = 1, 2, \dots, n.$$
(2.7)

Moreover, we denote $\overline{\Sigma} = \text{diag}\{\sigma_1^+, \dots, \sigma_n^+\}, \Sigma = \text{diag}\{\sigma_1^-, \dots, \sigma_n^-\}$, and

$$\Sigma_{1} = \text{diag}\{\sigma_{1}^{+}\sigma_{1}^{-}, \dots, \sigma_{n}^{+}\sigma_{n}^{-}\}, \qquad \Sigma_{2} = \text{diag}\left\{\frac{\sigma_{1}^{+} + \sigma_{1}^{-}}{2}, \dots, \frac{\sigma_{n}^{+} + \sigma_{n}^{-}}{2}\right\}.$$
 (2.8)

In the paper, we adopt the following definition.

Definition 2.1 (see [18]). For the system (2.6) and every initial condition $\varphi = \phi - \psi \in L^2_{\varphi}([-2\tau_{\max}, 0]; \mathbf{R}^n)$, the trivial solution is globally exponentially stable in the mean square, if there exist two positive scalars μ, k such that

$$\mathbf{E} \left\| \varepsilon(t; \varphi) \right\|^{2} \leq \mu \sup_{-\tau_{\max} \leq s \leq 0} \mathbf{E} \left\| \phi(s) - \psi(s) \right\|^{2} e^{-kt}, \quad \forall t \geq 0,$$
(2.9)

where **E** stands for the mathematical expectation and ϕ , ψ are the initial conditions of systems (2.1) and (2.2), respectively.

In many real applications, we are interested in designing a memoryless state-feedback controller $u(t) = K\varepsilon(t)$, where $K \in \mathbb{R}^{n \times n}$ is a constant gain matrix. In the paper, for a special case that the information on the size of $\tau(t)$ is available, we consider the delayed feedback controller of the following form:

$$u(t) = K\varepsilon(t) + K_1\varepsilon(t - \tau(t)), \qquad (2.10)$$

then replacing u(t) into system (2.6) yields

$$d\varepsilon(t) = \left[-\beta(\varepsilon(t)) + K\varepsilon(t) + K_1\varepsilon(t - \tau(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t))) + D\int_{t-\varrho(t)}^t f(\varepsilon(s))ds\right]dt + \sigma(t,\varepsilon(t),\varepsilon(t - \tau(t)))dw(t).$$
(2.11)

Then the purpose of the paper is to design a controller u(t) in (2.10) to let the slave system (2.2) synchronize with the master one (2.1).

3. Main Results

In this section, some lemmas are introduced firstly.

Lemma 3.1 (see [18]). For any symmetric matrix $W \in \mathbf{R}^{n \times n}$, $W = W^T \ge 0$, scalar h > 0, vector function $\omega : [0,h] \to \mathbf{R}^n$ such that the integrations concerned are well defined, then $(\int_0^h \omega(s)ds)^T W(\int_0^h \omega(s)ds) \le h \int_0^h \omega^T(s) W \omega(s) ds$.

Lemma 3.2 (see [19]). Given constant matrices P, Q, R, where $P^T = P$, $Q^T = Q$, then the linear matrix inequality (LMI) $\begin{bmatrix} P & R \\ R^T & -Q \end{bmatrix} < 0$ is equivalent to the condition: Q > 0, $P + RQ^{-1}R^T < 0$.

Lemma 3.3 (see [31]). Suppose that Ω , Ξ_{1i} , Ξ_{2i} , i = 1, 2 are the constant matrices of the appropriate dimensions, $\alpha \in [0, 1]$, and $\beta \in [0, 1]$, then the inequality $\Omega + [\alpha \Xi_{11} + (1-\alpha)\Xi_{12}] + [\beta \Xi_{21} + (1-\beta)\Xi_{22}] < 0$ holds, if the four inequalities $\Omega + \Xi_{11} + \Xi_{21} < 0$, $\Omega + \Xi_{11} + \Xi_{22} < 0$, $\Omega + \Xi_{12} + \Xi_{21} < 0$, $\Omega + \Xi_{12} + \Xi_{22} < 0$ hold simultaneously.

Then, a novel criterion is presented for the exponential stability for system (2.11) which can guarantee the master system (2.1) to synchronize the slave one (2.2).

Theorem 3.4. Supposing that assumptions (A1)–(A4) hold, then system (2.11) has one equilibrium point and is globally exponentially stable in the mean square, if there exist $n \times n$ matrices P > 0, $Q_j > 0$, $R_j > 0$ (j = 1, 2, 3), $Z_i > 0, S_i > 0, T_i > 0$, P_i (i = 1, 2), $n \times n$ diagonal matrices L > 0, Q > 0, H > 0, U > 0, V > 0, W > 0, R > 0, E > 0, $13n \times n$ matrices M, N, G, and one scalar $\lambda \ge 0$ such that the matrix inequalities (3.1)-(3.2) hold:

$$-\lambda I + P + (L+H)\left(\overline{\Sigma} - \Sigma\right) + Q(\Pi - \Gamma) + \overline{\tau}_m Z_2 + \tau_0 S_2 \le 0, \tag{3.1}$$

$$\begin{bmatrix} \Omega + \$ + \$^T - I_i T_2 I_i^T & \Xi_1 \\ * & \Phi \end{bmatrix} < 0, \qquad \begin{bmatrix} \Omega + \$ + \$^T - I_i T_2 I_i^T & \Xi_2 \\ * & \Phi \end{bmatrix} < 0, \quad i = 1, 2, \qquad (3.2)$$

where $I_1 = [0_{n \cdot 10n} \quad I_n \quad 0_{n \cdot 2n}], I_2 = [0_{n \cdot 11n} \quad I_n \quad 0_{n \cdot n}] and$

With

$$\begin{aligned} \Omega_{11} &= P_1^T K + K^T P_1 + Q_2 - U \Sigma_1 - 2\Gamma E + \lambda \Pi_1^T \Pi_1, \\ \Omega_{17} &= K^T P_2 + P - P_1^T + \overline{\Sigma} H - \Sigma L - \Gamma Q, \\ \Omega_{1,13} &= -P_1^T + K^T Q + E, \qquad \Omega_{22} = -W \Sigma_1 + Q_1 + Q_3 - Q_2, \\ \Omega_{33} &= -Q_3 - R \Sigma_1, \qquad \Omega_{44} = -U + R_2 + \varrho_0^2 T_1 + \overline{\varrho}_m^2 T_2, \end{aligned}$$
(3.4)
$$\begin{aligned} \Omega_{47} &= L - H + A^T P_2, \qquad \Omega_{55} = -W + R_1 + R_3 - R_2, \qquad \Omega_{66} = -R - R_3, \\ \Omega_{77} &= -P_2^T - P_2 + \overline{\tau}_m Z_1 + \tau_0 S_1, \qquad \Omega_{88} = -(1 - \mu)Q_1 - V \Sigma_1 + \lambda \Pi_2^T \Pi_2, \\ \Omega_{99} &= -(1 - \mu)R_1 - V. \end{aligned}$$

Proof. Denoting $\sigma(t) = \sigma(t, \varepsilon(t), \varepsilon(t - \tau(t)))$, we represent system (2.11) as the following equivalent form:

$$d\varepsilon(t) = v(t)dt + \sigma(t)dw(t),$$

$$v(t) = -\beta(\varepsilon(t)) + K\varepsilon(t) + K_1\varepsilon(t - \tau(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t)))$$

$$+ D\int_{t-\varrho(t)}^t f(\varepsilon(s))ds.$$
(3.5)

Now, together with assumptions (A1) and (A2), we construct the following Lyapunov-Krasovskii functional:

$$V(\varepsilon_t) = V_1(\varepsilon_t) + V_2(\varepsilon_t) + V_3(\varepsilon_t) + V_4(\varepsilon_t),$$
(3.6)

where

$$\begin{aligned} V_{1}(\varepsilon_{t}) &= \varepsilon^{T}(t)P\varepsilon(t) + 2\sum_{j=1}^{n} l_{j} \int_{0}^{\varepsilon_{j}} \left[f_{j}(s) - \sigma_{j}^{-}s\right] ds + 2\sum_{j=1}^{n} h_{j} \int_{0}^{\varepsilon_{j}} \left[\sigma_{j}^{+}s - f_{j}(s)\right] ds \\ &+ 2\sum_{j=1}^{n} q_{j} \int_{0}^{\varepsilon_{j}} \left[\beta_{j}(s) - \gamma_{j}s\right] ds, \end{aligned}$$

$$\begin{aligned} V_{2}(\varepsilon_{t}) &= \int_{t-\tau(t)}^{t-\tau_{0}} \left[\varepsilon^{T}(s)Q_{1}\varepsilon(s) + f^{T}(\varepsilon(s))R_{1}f(\varepsilon(s))\right] ds \\ &+ \int_{t-\tau_{0}}^{t} \left[\varepsilon^{T}(s)Q_{2}\varepsilon(s) + f^{T}(\varepsilon(s))R_{2}f(\varepsilon(s))\right] ds \\ &+ \int_{t-\tau_{m}}^{t-\tau_{0}} \left[\varepsilon^{T}(s)Q_{3}\varepsilon(s) + f^{T}(\varepsilon(s))R_{3}f(\varepsilon(s))\right] ds, \end{aligned}$$

$$\begin{aligned} V_{3}(\varepsilon_{t}) &= \int_{-\tau_{m}}^{-\tau_{0}} \int_{t+\theta}^{t} \nu^{T}(s)Z_{1}\nu(s)ds \, d\theta + \int_{-\tau_{m}}^{-\tau_{0}} \int_{t+\theta}^{t} \operatorname{trace}\left(\sigma^{T}(s)Z_{2}\sigma(s)\right) ds \, d\theta \\ &+ \int_{-\tau_{0}}^{0} \int_{t+\theta}^{t} \nu^{T}(s)S_{1}\nu(s)ds \, d\theta + \int_{-\tau_{0}}^{0} \int_{t+\theta}^{t} \operatorname{trace}\left(\sigma^{T}(s)S_{2}\sigma(s)\right) ds \, d\theta, \end{aligned}$$

$$\begin{aligned} V_{4}(\varepsilon_{t}) &= Q_{0} \int_{-Q_{0}}^{0} \int_{t+\theta}^{t} f^{T}(\varepsilon(s))T_{1}f(\varepsilon(s)) ds \, d\theta + \overline{Q}_{m} \int_{-Q_{m}}^{-Q_{0}} \int_{t+\theta}^{t} f^{T}(\varepsilon(s))T_{2}f(\varepsilon(s)) ds \, d\theta \end{aligned}$$

$$(3.7)$$

with setting $L = \text{diag}\{l_1, \dots, l_n\} > 0$, $H = \text{diag}\{h_1, \dots, h_n\} > 0$, and $Q = \text{diag}\{q_1, \dots, q_n\} > 0$. In the following, the weak infinitesimal operator \mathcal{L} of the stochastic process $\{\varepsilon_t, t \ge 0\}$ is given in [32].

By employing (A1) and (A2) and directly computing $\mathcal{L}V_i(\varepsilon_t)$ (i = 1, 2, 3, 4), it follows from any $n \times n$ matrices P_1, P_2 that

$$\begin{aligned} \mathcal{L}V_{1}(\varepsilon_{t}) \\ &\leq 2\varepsilon^{T}(t)P\nu(t) + 2\Big[f^{T}(\varepsilon(t)) - \varepsilon^{T}(t)\Sigma\Big]L\nu(t) + 2\Big[\varepsilon^{T}(t)\overline{\Sigma} - f^{T}(\varepsilon(t))\Big]H\nu(t) + 2\beta^{T}(\varepsilon(t))Q \\ &\times \left[-\beta(\varepsilon(t)) + K\varepsilon(t) + K_{1}\varepsilon(t - \tau(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t))) + D\int_{t-\varphi(t)}^{t} f(\varepsilon(s))ds\right] \\ &- 2\varepsilon^{T}(t)\Gamma^{T}Q\nu(t) + \operatorname{trace}\Big[\sigma^{T}(t)\Big[P + (L + H)\Big(\overline{\Sigma} - \Sigma\Big) + Q(\Pi - \Gamma)\Big]\sigma(t)\Big] + 2\Big[\varepsilon^{T}(t)P_{1}^{T} + \nu(t)P_{2}^{T}\Big] \end{aligned}$$

$$\times \left[-\nu(t) - \beta(\varepsilon(t)) + K\varepsilon(t) + K_1\varepsilon(t - \tau(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t))) + D\int_{t-\varrho(t)}^t f(\varepsilon(s))ds \right],$$
(3.8)

$$\mathcal{L}V_{2}(\varepsilon_{t})$$

$$\leq \left[\varepsilon^{T}(t-\tau_{0})Q_{1}\varepsilon(t-\tau_{0})+f^{T}(\varepsilon(t-\tau_{0}))R_{1}f(\varepsilon(t-\tau_{0}))\right]$$

$$-\left(1-\mu\right)\left[\varepsilon^{T}(t-\tau(t))Q_{1}\varepsilon(t-\tau(t))+f^{T}(\varepsilon(t-\tau(t)))R_{1}f(\varepsilon(t-\tau(t)))\right]$$

$$+\left[\varepsilon^{T}(t)Q_{2}\varepsilon(t)+f^{T}(\varepsilon(t))R_{2}f(\varepsilon(t))\right]$$

$$+\left[\varepsilon^{T}(t-\tau_{0})\times(Q_{3}-Q_{2})\varepsilon(t-\tau_{0})+f^{T}(\varepsilon(t-\tau_{0}))(R_{3}-R_{2})f(\varepsilon(t-\tau_{0}))\right]$$

$$-\left[\varepsilon^{T}(t-\tau_{m})Q_{3}\varepsilon(t-\tau_{m})+f^{T}(\varepsilon(t-\tau_{m}))R_{3}f(\varepsilon(t-\tau_{m}))\right],$$
(3.9)

$$\mathcal{L}V_3(\varepsilon_t)$$

$$=\overline{\tau}_{m}\nu^{T}(t)Z_{1}\nu(t) - \int_{t-\tau_{m}}^{t-\tau_{0}}\nu^{T}(s)Z_{1}\nu(s)ds + \overline{\tau}_{m}\operatorname{trace}\left[\sigma^{T}(t)Z_{2}\sigma(t)\right] - \int_{t-\tau_{m}}^{t-\tau_{0}}\operatorname{trace}\left[\sigma^{T}(s)Z_{2}\sigma(s)\right]ds + \tau_{0}\nu^{T}(t)S_{1}\nu(t) - \int_{t-\tau_{0}}^{t}\nu^{T}(s)S_{1}\nu(s)ds + \tau_{0}\operatorname{trace}\left[\sigma^{T}(t)S_{2}\sigma(t)\right] - \int_{t-\tau_{0}}^{t}\operatorname{trace}\left[\sigma^{T}(s)S_{2}\sigma(s)\right]ds,$$
(3.10)

$$\mathcal{L}V_{4}(\varepsilon_{t})$$

$$= f^{T}(\varepsilon(t)) \left(\varphi_{0}^{2}T_{1} + \overline{\varphi}_{m}^{2}T_{2} \right) f(\varepsilon(t)) - \int_{t-\varphi_{0}}^{t} \varphi_{0}f^{T}(\varepsilon(s))T_{1}f(\varepsilon(s))ds - \int_{t-\varphi_{m}}^{t-\varphi_{0}} \overline{\varphi}_{m}f^{T}(\varepsilon(s))T_{2}f(\varepsilon(s))ds$$

$$\leq f^{T}(\varepsilon(t)) \left(\varphi_{0}^{2}T_{1} + \overline{\varphi}_{m}^{2}T_{2} \right) f(\varepsilon(t)) - \left[\int_{t-\varphi_{0}}^{t} f(\varepsilon(s))ds \right]^{T} T_{1} \left[\int_{t-\varphi_{0}}^{t} f(\varepsilon(s))ds \right] - (1 + \mu_{1})$$

$$\times \left[\int_{t-\varphi(t)}^{t-\varphi_{0}} f(\varepsilon(s))ds \right]^{T} T_{2} \left[\int_{t-\varphi(t)}^{t-\varphi_{0}} f(\varepsilon(s))ds \right]$$

$$- (1 + \mu_{2}) \left[\int_{t-\varphi_{m}}^{t-\varphi(t)} f^{T}(\varepsilon(s))ds \right] T_{2} \left[\int_{t-\varphi_{m}}^{t-\varphi(t)} f(\varepsilon(s))ds \right], \qquad (3.11)$$

where $\mu_1 = (\rho_m - \rho(t))/\overline{\rho}_m$ and $\mu_2 = (\rho(t) - \rho_0)/\overline{\rho}_m$.

Now adding the terms on the right side of (3.8)–(3.11) to $\mathcal{L}V(\varepsilon_t)$ and employing (2.5), (3.1), it is easy to obtain

$$\begin{split} \mathcal{L} V(\varepsilon_{l}) &\leq 2\varepsilon^{T}(t) \mathcal{P} v(t) + 2 \Big[f^{T}(\varepsilon(t)) - \varepsilon^{T}(t) \Sigma \Big] L v(t) + 2 \Big[\varepsilon^{T}(t) \overline{\Sigma} - f^{T}(\varepsilon(t)) \Big] H v(t) \\ &+ 2 \Big[\varepsilon^{T}(t) \mathcal{P}_{l}^{T} + v(t) \mathcal{P}_{2}^{T} + \beta^{T}(\varepsilon(t)) Q \Big] \\ &\times \Bigg[-\beta(\varepsilon(t)) + K\varepsilon(t) + K_{1}\varepsilon(t - \tau(t)) + Af(\varepsilon(t)) + Bf(\varepsilon(t - \tau(t))) \\ &+ D \bigg[\int_{t-\sqrt{0}}^{t} f(\varepsilon(s)) ds + \int_{t-\sqrt{0}}^{t-\sqrt{0}} f(\varepsilon(s)) ds \Big] \Bigg] \\ &- 2 \Big[\varepsilon^{T}(t) \mathcal{P}_{1}^{T} + v(t) \mathcal{P}_{2}^{T} \Big] v(t) - 2\varepsilon^{T}(t) \Gamma^{T} Q v(t) \\ &+ \Big[\varepsilon^{T}(t - \tau_{0}) Q_{1}\varepsilon(t - \tau_{0}) + f^{T}(\varepsilon(t - \tau_{0})) R_{1}f(\varepsilon(t - \tau_{0})) \Big] \\ &- (1 - \mu) \Big[\varepsilon^{T}(t - \tau(t)) Q_{1}\varepsilon(t - \tau(t)) + f^{T}(\varepsilon(t - \tau_{0})) R_{1}f(\varepsilon(t - \tau_{0})) \Big] \\ &+ \Big[\varepsilon^{T}(t) Q_{2}\varepsilon(t) + f^{T}(\varepsilon(t)) R_{2}f(\varepsilon(t)) \Big] \\ &+ \Big[\varepsilon^{T}(t - \tau_{0}) (Q_{3} - Q_{2})\varepsilon(t - \tau_{0}) + f^{T}(\varepsilon(t - \tau_{0})) (R_{3} - R_{2})f(\varepsilon(t - \tau_{0})) \Big] \\ &- \Big[\varepsilon^{T}(t - \tau_{m}) Q_{3}\varepsilon(t - \tau_{m}) + f^{T}(\varepsilon(t - \tau_{m})) R_{3}f(\varepsilon(t - \tau_{m})) \Big] + v^{T}(t)(\overline{\tau}_{m} Z_{1} + \tau_{0} S_{1}) v(t) \\ &- \int_{t-\tau_{0}}^{t-\tau_{0}} v^{T}(s) Z_{1} v(s) ds - \int_{t-\tau_{0}}^{t} v^{T}(s) S_{1} v(s) ds - \int_{t-\tau_{m}}^{t-\tau_{0}} t \operatorname{trace} \Big[\sigma^{T}(s) Z_{2} \sigma(s) \Big] ds \\ &- \int_{t-\tau_{0}}^{t} \operatorname{trace} \Big[\sigma^{T}(s) S_{2} \sigma(s) \Big] ds + \lambda \Big[\varepsilon^{T}(t) \Pi_{1}^{T} \Pi_{1}\varepsilon(t) + \varepsilon^{T}(t - \tau(t)) \Pi_{2}^{T} \Pi_{2}\varepsilon(t - \tau(t)) \Big] \\ &+ f^{T}(\varepsilon(t)) \Big(q_{0}^{2} T_{1} + \overline{q}_{m}^{2} T_{2} \Big) f(\varepsilon(t)) - \Big[\int_{t-q_{0}}^{t-q_{0}} f(\varepsilon(s)) ds \Big]^{T} T_{1} \Big[\int_{t-q_{0}}^{t} f(\varepsilon(s)) ds \Big] - (1 + \mu_{1}) \Big[\int_{t-q_{0}}^{t-q_{0}} f^{T}(\varepsilon(s)) ds \Big] T_{2} \Big[\int_{t-q_{0}}^{t-q_{0}} f(\varepsilon(s)) ds \Big] . \end{split}$$

(3.12)

Based on methods in [33] and (2.7), for any $n \times n$ diagonal matrices U > 0, V > 0, W > 0, R > 0, the following inequality can be achieved:

$$0 \leq -\left[x^{T}(t)U\Sigma_{1}x(t) - 2x^{T}(t)U\Sigma_{2}f(x(t)) + f^{T}(x(t))Uf(x(t))\right] -\left[x^{T}(t-\tau(t))V\Sigma_{1}x(t-\tau(t)) - 2x^{T}(t-\tau(t))V\Sigma_{2}f(x(t-\tau(t))) + f^{T}(x(t-\tau(t)))Vf(x(t-\tau(t)))\right] -\left[x^{T}(t-\tau_{0})W\Sigma_{1}x(t-\tau_{0}) - 2x^{T}(t-\tau_{0})W\Sigma_{2}f(x(t-\tau_{0})) + f^{T}(x(t-\tau_{0}))Wf(x(t-\tau_{0}))\right] -\left[x^{T}(t-\tau_{m})R\Sigma_{1}x(t-\tau_{m}) - 2x^{T}(t-\tau_{m})R\Sigma_{2}f(x(t-\tau_{m})) + f^{T}(x(t-\tau_{m}))Rf(x(t-\tau_{m}))\right].$$
(3.13)

From (A1), for any $n \times n$ diagonal matrix *E*, one can yield

$$0 \le 2 \left[\beta(\varepsilon(t)) - \Gamma \varepsilon(t) \right]^T E \varepsilon(t). \tag{3.14}$$

Furthermore, for any $13n \times n$ constant matrices *M*, *N*, *G*, we can obtain

$$0 = 2\zeta^{T}(t)M\left[\varepsilon(t) - \varepsilon(t - \tau_{0}) - \int_{t-\tau_{0}}^{t} v(s)ds - \int_{t-\tau_{0}}^{t} \sigma(s)d\omega(s)\right]$$

+ $2\zeta^{T}(t)N\left[\varepsilon(t - \tau_{0}) - \varepsilon(t - \tau(t)) - \int_{t-\tau(t)}^{t-\tau_{0}} v(s)ds - \int_{t-\tau(t)}^{t-\tau_{0}} \sigma(s)d\omega(s)\right]$
+ $2\zeta^{T}(t)G\left[\varepsilon(t - \tau(t)) - \varepsilon(t - \tau_{m}) - \int_{t-\tau_{m}}^{t-\tau(t)} v(s)ds - \int_{t-\tau_{m}}^{t-\tau(t)} \sigma(s)d\omega(s)\right],$ (3.15)

where

$$\zeta^{T}(t) = \left[\varepsilon^{T}(t)\varepsilon^{T}(t-\tau_{0})\varepsilon^{T}(t-\tau_{m})f^{T}(\varepsilon(t))f^{T}(\varepsilon(t-\tau_{0}))f^{T}(\varepsilon(t-\tau_{m})) \right]$$
$$v^{T}(t)\varepsilon^{T}(t-\tau(t))f^{T}(\varepsilon(t-\tau(t)))\left[\int_{t-\varrho_{0}}^{t}f(\varepsilon(s)ds)\right]^{T} \left[\int_{t-\varrho_{0}}^{t-\varrho_{0}}f(\varepsilon(s)ds)\right]^{T} \left[\int_{t-\varrho_{m}}^{t-\varrho(t)}f(\varepsilon(s)ds)\right]^{T} \beta^{T}(\varepsilon(t))\right].$$
(3.16)

Then together with the methods in [28, 29], combining (3.12)-(3.15) yields

$$\mathcal{L}V(\varepsilon_{t}) \leq \zeta^{T}(t) \Big[\Omega + \$ + \$^{T} + \tau_{0}MS_{1}^{-1}M^{T} + [\tau(t) - \tau_{0}]NZ_{1}^{-1}N^{T} + [\tau_{m} - \tau(t)]GZ_{1}^{-1}G^{T} -\mu_{1}I_{1}T_{2}I_{1}^{T} - \mu_{2}I_{2}T_{2}I_{2}^{T} + MS_{2}^{-1}M^{T} + NZ_{2}^{-1}N^{T} + GZ_{2}^{-1}G^{T} \Big]$$

$$\times \zeta(t) + h(t) := \zeta^{T}(t)\Delta(t)\zeta(t) + h(t),$$
(3.17)

where Ω , \$ are presented in (3.2) and

$$h(t) = \left[\int_{t-\tau_0}^t \sigma(s)d\omega(s)\right]^T S_2 \left[\int_{t-\tau_0}^t \sigma(s)d\omega(s)\right] + \left[\int_{t-\tau(t)}^{t-\tau_0} \sigma(s)d\omega(s)\right]^T Z_2 \left[\int_{t-\tau(t)}^{t-\tau_0} \sigma(s)d\omega(s)\right]$$
$$+ \left[\int_{t-\tau_m}^{t-\tau(t)} \sigma(s)d\omega(s)\right]^T Z_2 \left[\int_{t-\tau_m}^{t-\tau(t)} \sigma(s)d\omega(s)\right] - \int_{t-\tau_0}^t \operatorname{trace}\left[\sigma^T(s)S_2\sigma(s)\right] ds$$
$$- \int_{t-\tau(t)}^{t-\tau_0} \operatorname{trace}\left[\sigma^T(s)Z_2\sigma(s)\right] ds - \int_{t-\tau_m}^{t-\tau(t)} \operatorname{trace}\left[\sigma^T(s)Z_2\sigma(s)\right] ds.$$
(3.18)

Together with Lemmas 3.2 and 3.3, the nonlinear matrix inequalities in (3.2) can guarantee $\Delta(t) < 0$ to be true. Therefore, there must exist a negative scalar $\chi < 0$ such that

$$\mathcal{L}V(\varepsilon_t) \le \zeta^T(t)\Delta(t)\zeta(t) + h(t) \le \chi \Big[\|\varepsilon(t)\|^2 + \|\varepsilon(t-\tau(t))\|^2 \Big] + h(t).$$
(3.19)

Taking the mathematic expectation of (3.19), we can deduce $Eh(t) = 0, E\mathcal{L}V(\varepsilon_t) \leq \chi E[\|\varepsilon(t)\|^2 + \|\varepsilon(t - \tau(t))\|^2]$, which indicates that the dynamics of the system (2.11) is globally asymptotically stable in the mean square. Based on $V(\varepsilon_t)$ in (3.6) and directly computing, there must exist three positive scalars $\Theta_i > 0$, i = 1, 2, 3 such that

$$V(\varepsilon_t) \le \Theta_1 \|\varepsilon(t)\|^2 + \Theta_2 \int_{t-\tau_{\max}}^t \|\varepsilon(v)\|^2 dv + \Theta_3 \int_{t-\tau_m}^t \|\varepsilon(v-\tau(v))\|^2 dv.$$
(3.20)

Letting $\overline{V}(\varepsilon_t) = e^{kt}V(\varepsilon_t)$, we can deduce

$$\begin{split} \mathbf{E}\overline{V}(\varepsilon_{t}) - \mathbf{E}\overline{V}(\varepsilon_{0}) &= \mathbf{E}\int_{0}^{t} \mathcal{L}\left(e^{ks}V(\varepsilon_{s})\right) ds \\ &\leq \mathbf{E}\int_{0}^{t} e^{ks} \left\{ k \left[\Theta_{1} \|\varepsilon(s)\|^{2} + \Theta_{2} \int_{s-\tau_{\max}}^{s} \|\varepsilon(v)\|^{2} dv + \Theta_{3} \int_{s-\tau_{m}}^{s} \|\varepsilon(v-\tau(v))\|^{2} dv \right] \\ &+ \chi \left[\|\varepsilon(s)\|^{2} + \|\varepsilon(s-\tau(s))\|^{2} \right] \right\} ds. \end{split}$$

$$(3.21)$$

By changing the integration sequence, it can be deduced that

$$\int_{0}^{t} e^{ks} \int_{s-\tau_{\max}}^{s} \|\varepsilon(v)\|^{2} dv ds$$

$$\leq \int_{-\tau_{\max}}^{t} \int_{v}^{v+\tau_{\max}} e^{ks} \|\varepsilon(v)\|^{2} ds dv$$

$$\leq \tau_{\max} e^{k\tau_{\max}} \int_{-\tau_{\max}}^{t} \|\varepsilon(v)\|^{2} e^{kv} dv,$$

$$\int_{0}^{t} e^{ks} \int_{s-\tau_{m}}^{s} \|\varepsilon(v-\tau(v))\|^{2} dv ds \leq \tau_{m} e^{k\tau_{m}} \int_{-\tau_{m}}^{t} \|\varepsilon(v-\tau(v))\|^{2} e^{kv} dv.$$
(3.22)

Substituting the terms (3.22) into the relevant ones in (3.21), it is easy to have

$$\mathbf{E}\overline{V}(\varepsilon_{t}) \leq \mathbf{E}\overline{V}(\varepsilon_{0}) + \mathbf{E}\left\{\left[k\Theta_{1} + k\Theta_{2}\tau_{\max}e^{k\tau_{\max}} + \chi\right]\int_{0}^{t}\|\varepsilon(v)\|^{2}e^{kv}dv + \left[k\Theta_{3}\tau_{m}e^{k\tau_{m}} + \chi\right]\int_{0}^{t}\|\varepsilon(v-\tau(v))\|^{2}e^{kv}dv + h_{0}(k)\right\},$$
(3.23)

where $h_0(k) = k\Theta_2 \tau_{\max} e^{k\tau_{\max}} \int_{-\tau_{\max}}^0 \|\varepsilon(v)\|^2 e^{kv} dv + k\Theta_3 \tau_m e^{k\tau_m} \int_{-\tau_m}^0 \|\varepsilon(v - \tau(v))\|^2 e^{kv} dv$. Choose one sufficiently small scalar $k_0 > 0$ such that $k_0\Theta_1 + k_0\Theta_2 \tau_{\max} e^{k_0\tau_{\max}} + \chi \le 0$, $k_0\Theta_3 \tau_m e^{k_0\tau_m} + \chi \le 0$. Then, $\mathbf{E}\overline{V}(\varepsilon_t) \le \mathbf{E}h_0(k_0) + \mathbf{E}\overline{V}(\varepsilon_0)$. Through directly computing, there must exist a positive scalar $\Upsilon > 0$ such that

$$\mathbf{E}\overline{\mathbf{V}}(\varepsilon_0) + \mathbf{E}h_0(k_0) \le \Upsilon \sup_{-2\tau_{\max} \le s \le 0} \mathbf{E} \|\varphi(s)\|^2.$$
(3.24)

Meanwhile, $\mathbf{E}\overline{V}(\varepsilon_t) \ge \lambda_{\min}(P)e^{k_0t}\mathbf{E}\|\varepsilon(t)\|^2$. Thus with (3.24), one can obtain

$$\mathbf{E} \|\varepsilon(t)\|^2 \le \lambda_{\min}^{-1}(P) \Upsilon \sup_{-2\tau_{\max} \le s \le 0} \mathbf{E} \|\varphi(s)\|^2 e^{-k_0 t}, \quad \forall t \ge 0,$$
(3.25)

which indicates that system (2.11) is globally exponentially stable in the mean square, and the proof is completed. $\hfill \Box$

Remark 3.5. As for systems (2.1) and (2.2), many present literatures have much attention to $\beta(z(t)) = Cz(t)$ with *C* positive-definite diagonal matrix, which can be checked as one special case of assumption (A3). Also in Theorem 3.4, it can be checked that $\Delta(t) < 0$ in (3.17) was not simply enlarged by $\Omega + \$ + \$^T + \tau_0 M S_1^{-1} M^T + \overline{\tau}_m N Z_1^{-1} N^T + \overline{\tau}_m G Z_1^{-1} G^T + M S_2^{-1} M^T + N Z_2^{-1} N^T + G Z_2^{-1} G^T < 0$, but equivalently guaranteed by utilizing two matrix inequalities (3.2) and Lemma 3.3, which can be more effective than these techniques employed in [18, 28, 29]. Moreover, we compute and estimate $\mathcal{L}V_5(\varepsilon_t)$ in (3.11) more efficiently than those present ones owing to that some previously ignored terms have been taken into consideration.

In order to show the design of the estimate gain matrices K and K_1 , a simple transformation is made to obtain the following theorem.

Theorem 3.6. Supposing that assumptions (A1)–(A4) hold and setting $\epsilon_1, \epsilon_2 > 0$, then the system (2.1) and system (2.2) can exponentially achieve the master-slave synchronization in the mean square, if there exist $n \times n$ matrices $P > 0, Q_j > 0, R_j > 0$ (j = 1, 2, 3), $Z_i > 0, S_i > 0$, $T_i > 0$ (i = 1, 2), $F, F_1n \times n$ diagonal matrices $L > 0, Q > 0, H > 0, U > 0, V > 0, W > 0, R > 0, E > 0, 13n \times n$ matrices M, N, G, and one scalar $\lambda \ge 0$ such that the LMIs in (3.26)-(3.27) hold

$$-\lambda I + P + (L+H)\left(\overline{\Sigma} - \Sigma\right) + Q(\Pi - \Gamma) + \overline{\tau}_m Z_2 + \tau_0 S_2 \le 0, \tag{3.26}$$

$$\begin{bmatrix} \Xi + \$ + \$^T - I_i T_2 I_i^T & \Xi_1 \\ * & \Phi \end{bmatrix} < 0, \qquad \begin{bmatrix} \Xi + \$ + \$^T - I_i T_2 I_i^T & \Xi_2 \\ * & \Phi \end{bmatrix} < 0, \quad i = 1, 2,$$
(3.27)

where I_i , Ξ_i (i = 1, 2), \mathcal{G} , Φ are similar to the relevant ones in (3.2), and

	Ξ_{11}	0	0	Ξ_{14}	0	0	Ξ_{17}	$\epsilon_1 F_1$	$\epsilon_1 QB$	$\epsilon_1 QD$	$\epsilon_1 QD$	0	Ξ _{1,13}	
	*	Ξ_{22}	0	0	$W\Sigma_2$	0	0	0	0	0	0	0	0	
	*	*	Ξ_{33}	0	0	$R\Sigma_2$	0	0	0	0	0	0	0	
	*	*	*	Ξ_{44}	0	0	Ξ_{47}	0	0	0	0	0	$A^T Q$	
	*	*	*	*	Ξ_{55}	0	0	0	0	0	0	0	0	
	*	*	*	*	*	$-R - R_3$	0	0	0	0	0	0	0	
Ξ=	*	*	*	*	*	*	Ξ77	$\epsilon_2 F_1$	$\epsilon_2 QB$	$\epsilon_2 QD$	$\epsilon_2 QD$	0	$-\epsilon_2 Q$,
	*	*	*	*	*	*	*	Ξ_{88}	$V\Sigma_2$	0	0	0	F_1^T	
	*	*	*	*	*	*	*	*	Ξ ₉₉	0	0	0	$B^T Q$	
	*	*	*	*	*	*	*	*	*	$-T_1$	0	0	$D^T Q$	
	*	*	*	*	*	*	*	*	*	*	$-T_2$	0	$D^T Q$	
	*	*	*	*	*	*	*	*	*	*	*	$-T_2$	0	
	*	*	*	*	*	*	*	*	*	*	*	*	$-Q - Q^T$	
_													(3	3.28)

with

$$\Xi_{11} = \epsilon_{1}F + \epsilon_{1}F^{T} + Q_{2} - U\Sigma_{1} - 2\Gamma E + \lambda\Pi_{1}^{T}\Pi_{1}, \qquad \Xi_{14} = \epsilon_{1}QA + U\Sigma_{2},$$

$$\Xi_{17} = \epsilon_{2}F^{T} + P - \epsilon_{1}Q + \overline{\Sigma}H - \Sigma L - \Gamma Q, \qquad \Xi_{1,13} = -\epsilon_{1}Q + F^{T} + E,$$

$$\Xi_{22} = -W\Sigma_{1} + Q_{1} + Q_{3} - Q_{2},$$

$$\Xi_{33} = -Q_{3} - R\Sigma_{1}, \qquad \Xi_{44} = -U + R_{2} + \rho_{0}^{2}T_{1} + \overline{\rho}_{m}^{2}T_{2}, \qquad (3.29)$$

$$\Xi_{47} = L - H + \epsilon_{2}A^{T}Q, \qquad \Xi_{55} = -W + R_{1} + R_{3} - R_{2},$$

$$\Xi_{77} = -\epsilon_{2}Q - \epsilon_{2}Q + \overline{\tau}_{m}Z_{1} + \tau_{0}S_{1}, \qquad \Xi_{88} = -(1 - \mu)Q_{1} - V\Sigma_{1} + \lambda\Pi_{2}^{T}\Pi_{2},$$

$$\Xi_{99} = -(1 - \mu)R_{1} - V.$$

Moreover, the estimation gains $K = Q^{-T}F$ *and* $K_1 = Q^{-T}F_1$.

Proof. Letting $P_1 = \epsilon_1 Q$, $P_2 = \epsilon_2 Q$ and setting $F = Q^T K$, $F_1 = Q^T K_1$ in (3.2) of Theorem 3.4, it is easy to derive the result and the detailed proof is omitted here.

Remark 3.7. Theorem 3.6 presents one novel delay-dependent criterion guaranteeing the systems (2.1) and (2.2) to achieve the master-slave synchronization in an exponential way. The method is presented in terms of LMIs, therefore, by using LMI in MATLAB Toolbox, it is straightforward and convenient to check the feasibility of the proposed results without tuning any parameters. Moreover, the systems addressed in this paper can include some famous networks in [17, 19–21, 23] as its special cases or $\tau(t)$ is not differentiable.

Remark 3.8. Through setting $Q_1 = R_1 = 0$ in (3.6) and employing similar methods, Theorems 3.4 and 3.6 can be applicable without taking the upper bound on derivative of $\tau(t)$ into consideration, which means that Theorems 3.4 and 3.6 can be true even as μ is unknown.

Remark 3.9. As we all know, most of $n \times n$ free-weighting matrices of M, N, G in Theorems 3.4 and 3.6 cannot help reduce the conservatism but only result in computational complexity. Thus we can choose the simplified slack matrices M, N, G as follows:

$$M = \begin{bmatrix} M_1 & M_2 & 0_{n \cdot 11n} \end{bmatrix}^T, \qquad N = \begin{bmatrix} 0_{n \cdot n} & N_1 & 0_{n \cdot 5n} & N_2 & 0_{n \cdot 5n} \end{bmatrix}^T, G = \begin{bmatrix} 0_{n \cdot 2n} & G_1 & 0_{n \cdot 4n} & G_2 & 0_{n \cdot 5n} \end{bmatrix}^T,$$
(3.30)

with $n \times n$ matrices M_i , N_i , G_i (i = 1, 2). Though the number of $n \times n$ matrix variables in (3.30) is much smaller than the one in (3.2) and (3.27), the numerical examples given in the paper still demonstrate that the simplified criteria can reduce the conservatism as effectively as Theorems 3.4 and 3.6 do.

4. Numerical Examples

In this section, two numerical examples will be given to illustrate the effectiveness of the proposed results.

Example 4.1. Consider the drive system (2.1) and response one (2.2) of delayed neural networks as follows:

$$b(z) = \begin{bmatrix} 0.7z_1 + 0.5 \tanh(z_1) \\ 0.7z_2 + 0.5 \tanh(z_2) \end{bmatrix}, \quad A = \begin{bmatrix} 0.2 & -0.4 \\ -0.4 & 0.2 \end{bmatrix}, \quad B = \begin{bmatrix} 0.2 & 0.2 \\ 0.2 & 0.2 \end{bmatrix}, \quad D = \begin{bmatrix} 0.2 & 0.3 \\ 0.1 & 0.21 \end{bmatrix},$$
$$g(z) = \begin{bmatrix} 0.25(|z_1 + 1| - |z_1 - 1|) \\ 0.25(|z_2 + 1| - |z_2 - 1|) \end{bmatrix}, \quad \sigma(t, \varepsilon(t), \varepsilon(t - \tau(t))) = 0.1 \times \begin{bmatrix} \|\varepsilon(t)\| & 0 \\ 0 & \|\varepsilon(t - \tau(t))\| \end{bmatrix},$$
$$\tau(t) = 1.0\sin^2(10t) + 0.5, \quad Q(t) = 2\cos^2 t + 0.5.$$
(4.1)

Then it is easy to check that $\tau_0 = 0.5$, $\tau_m = 1.5$, $\rho_0 = 0.5$, $\rho_m = 2.5$, $\mu = 10$, and

$$\Gamma = \begin{bmatrix} 0.7 & 0 \\ 0 & 0.7 \end{bmatrix}, \qquad \Pi = \begin{bmatrix} 1.2 & 0 \\ 0 & 1.2 \end{bmatrix}, \qquad \Sigma = \begin{bmatrix} -0.5 & 0 \\ 0 & -0.5 \end{bmatrix}, \qquad \overline{\Sigma} = \begin{bmatrix} 0.5 & 0 \\ 0 & 0.5 \end{bmatrix}, \qquad (4.2)$$
$$\Pi_1 = \Pi_2 = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}.$$

By setting $e_1 = 0.05$, $e_2 = 0.01$ and utilizing Theorem 3.6, then the estimator gain matrices *K* and K_1 in (2.10) can be worked out

$$K = Q^{-T}F = \begin{bmatrix} -2.3981 & -0.0575\\ -0.0575 & -2.3994 \end{bmatrix}, \qquad K_1 = Q^{-T}F_1 = \begin{bmatrix} -0.2175 & 0.3347\\ 0.3347 & -0.2175 \end{bmatrix}.$$
(4.3)

Furthermore, as for $\tau(t) = |\sin(20t)| + 0.5$, $\rho(t) = 2|\cos(6t)| + 0.5$, and setting $\epsilon_1 = 0.05$, $\epsilon_2 = 0.01$, we can obtain the following estimator gain matrices by using Theorem 3.6 and Remark 3.8:

$$K = Q^{-T}F = \begin{bmatrix} -2.5374 & -0.0528\\ -0.0528 & -2.5385 \end{bmatrix}, \qquad K_1 = Q^{-T}F_1 = \begin{bmatrix} -0.3021 & 0.3812\\ 0.3812 & -0.3021 \end{bmatrix},$$
(4.4)

which means that the obtained results still hold as the time delay is not differentiable. However, the methods proposed in [17–19] fail to solve the synchronization problem even without the distributed delay.

Example 4.2. As a special case, we consider the master system (2.1) of delayed stochastic neural networks as follows:

$$dz(t) = \left[-Cz(t) + Ag(z(t)) + Bg(z(t - \tau(t))) + D \int_{t-\varphi(t)}^{t} g(z(s))ds + I \right] dt,$$
(4.5)

where $C = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$, $A = \begin{bmatrix} 1.8 & -0.3 \\ -5.1 & 2.6 \end{bmatrix}$, $B = \begin{bmatrix} -1.6 & -0.1 \\ -0.3 & -2.5 \end{bmatrix}$, $D = \begin{bmatrix} 2 & 1 \\ 1 & 2 \end{bmatrix}$, $I = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$, and $\tau(t) = 0.95 + 0.05 \sin^2(40t)$, Q(t) = 0.1. It can be verified that $\tau_0 = 0.95$, $\tau_m = 1.0$, $\mu = 2$, and $Q_0 = Q_m = 0.1$. The activation functions can be taken as $g_i(s) = \tanh(s)$, $s \in \mathbb{R}$ (i = 1, 2). The corresponding slave system can be

$$dy(t) = \left[-Cy(t) + Ag(y(t)) + Bg(y(t - \tau(t))) + D \int_{t-\varrho(t)}^{t} g(y(s))ds + I + u(t) \right] dt$$

$$+ \sigma(t, \varepsilon(t), \varepsilon(t - \tau(t)))d\omega(t),$$
(4.6)



Figure 1: Phase trajectories and state trajectories of drive system, response system and error system.

where $\sigma(t, \varepsilon(t), \varepsilon(t - \tau(t))) = \begin{bmatrix} \|\varepsilon(t)\| & 0\\ 0 & \|\varepsilon(t - \tau(t))\| \end{bmatrix}$. Then together with Theorem 3.6, $\varepsilon_1 = 0.05$, and $\varepsilon_2 = 0.1$, we can obtain part feasible solution to the LMIs in (3.26) and (3.27) by resorting to the Matlab LMI Toolbox:

$$Q = \begin{bmatrix} 0.2526 & 0\\ 0 & 0.2526 \end{bmatrix}, \qquad F = \begin{bmatrix} -4.0139 & -0.0421\\ 0.0486 & -4.0182 \end{bmatrix}, \qquad F_1 = \begin{bmatrix} 0.1133 & 0.0043\\ 0.0168 & 0.1707 \end{bmatrix}.$$
(4.7)

Then the estimator gain matrices K, K_1 can be deduced as follows:

$$K = Q^{-T}F = \begin{bmatrix} -15.8892 & -0.1667\\ 0.1923 & -15.9062 \end{bmatrix}, \qquad K_1 = Q^{-T}F_1 = \begin{bmatrix} 0.4486 & 0.0170\\ 0.0666 & 0.6758 \end{bmatrix}.$$
(4.8)

It follows from Theorem 3.6 that the drive system with the initial condition $[0.5, 0.4]^T$ for $-1 \le t \le 0$ synchronizes with the response system when the initial condition is $[0.7, 0.6]^T$ for

 $-1 \le t \le 0$. The phase trajectories and state ones of drive system and response one and state trajectories of error system are shown in Figure 1. Therefore, from Figure 1, we can see that the master system synchronizes with the slave system.

5. Conclusions

In this paper, we consider the synchronization control of stochastic neural networks with both time-varying and distributed time-varying delays. By using the Lyapunov functional and LMI technique, one sufficient condition has been derived to ensure the global exponential stability for the error system, and thus, the slave system can synchronize the master one. Then, the estimation gains can be obtained. The obtained results are novel since the addressed networks are of more general forms and some good mathematical techniques are employed. Finally, we give two numerical examples to verify the theoretical results.

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References

- L. M. Pecora and T. L. Carroll, "Synchronization in chaotic systems," *Physical Review Letters*, vol. 64, no. 8, pp. 821–824, 1990.
- [2] T. L. Liao and S. H. Tsai, "Adaptive synchronization of chaotic systems and its application to secure communications," *Chaos, Solitons & Fractals*, vol. 11, no. 9, pp. 1387–1396, 2000.
- [3] V. Perez-Munuzuri, V. Perez-Villar, and L. O. Chua, "Autowaves for image processing on a two-dimensional CNN array of excitable nonlinear circuits: flat and wrinkled labyrinths," *IEEE Transactions on Circuits and Systems I*, vol. 40, no. 3, pp. 174–181, 1993.
- [4] F. Zou and J. A. Nossek, "Bifurcation and chaos in cellular neural networks," IEEE Transactions on Circuits and Systems I, vol. 40, no. 3, pp. 166–173, 1993.
- [5] M. Gilli, "Strange attractors in delayed cellular neural networks," *IEEE Transactions on Circuits and Systems I*, vol. 40, no. 11, pp. 849–853, 1993.
- [6] C.-J. Cheng, T.-L. Liao, and C.-C. Hwang, "Exponential synchronization of a class of chaotic neural networks," *Chaos, Solitons & Fractals*, vol. 24, no. 1, pp. 197–206, 2005.
- [7] J.-J. Yan, J.-S. Lin, M.-L. Hung, and T.-L. Liao, "On the synchronization of neural networks containing time-varying delays and sector nonlinearity," *Physics Letters A*, vol. 361, no. 1-2, pp. 70–77, 2007.
- [8] H. Huang and G. Feng, "Synchronization of nonidentical chaotic neural networks with time delays," *Neural Networks*, vol. 22, no. 7, pp. 869–874, 2009.
- [9] Q. Song, "Design of controller on synchronization of chaotic neural networks with mixed timevarying delays," *Neurocomputing*, vol. 72, no. 13–15, pp. 3288–3295, 2009.
- [10] T. Li, S. M. Fei, Q. Zhu, and S. Cong, "Exponential synchronization of chaotic neural networks with mixed delays," *Neurocomputing*, vol. 71, no. 13–15, pp. 3005–3019, 2008.
- [11] T. Li, A. G. Song, S. M. Fei, and Y. Q. Guo, "Synchronization control of chaotic neural networks with time-varying and distributed delays," *Nonlinear Analysis, Theory, Methods & Applications*, vol. 71, no. 5-6, pp. 2372–2384, 2009.
- [12] J. H. Park, "Synchronization of cellular neural networks of neutral type via dynamic feedback controller," Chaos, Solitons & Fractals, vol. 42, no. 3, pp. 1299–1304, 2009.
- [13] H. Li and D. Yue, "Synchronization stability of general complex dynamical networks with timevarying delays: a piecewise analysis method," *Journal of Computational and Applied Mathematics*, vol. 232, no. 2, pp. 149–158, 2009.
- [14] Y. Liu, Z. Wang, J. Liang, and X. Liu, "Synchronization and state estimation for discrete-time complex networks with distributed delays," *IEEE Transactions on Systems, Man, and Cybernetics Part B*, vol. 38, no. 5, pp. 1314–1325, 2008.

- [15] H. Zhang, Y. Xie, Z. Wang, and C. Zheng, "Adaptive synchronization between two different chaotic neural networks with time delay," *IEEE Transactions on Neural Networks*, vol. 18, no. 6, pp. 1841–1845, 2007.
- [16] H. R. Karimi and P. Maass, "Delay-range-dependent exponential H_∞ synchronization of a class of delayed neural networks," *Chaos, Solitons & Fractals*, vol. 41, no. 3, pp. 1125–1135, 2009.
- [17] W. Yu and J. Cao, "Synchronization control of stochastic delayed neural networks," *Physica A*, vol. 373, pp. 252–260, 2007.
- [18] Y. Tang, J. A. Fang, and Q. Miao, "On the exponential synchronization of stochastic jumping chaotic neural networks with mixed delays and sector-bounded non-linearities," *Neurocomputing*, vol. 72, no. 7-9, pp. 1694–1701, 2009.
- [19] J. H. Park and O. M. Kwon, "Synchronization of neural networks of neutral type with stochastic perturbation," *Modern Physics Letters B*, vol. 23, no. 14, pp. 1743–1751, 2009.
- [20] X. Li and J. Cao, "Adaptive synchronization for delayed neural networks with stochastic perturbation," *Journal of the Franklin Institute*, vol. 345, no. 7, pp. 779–791, 2008.
- [21] Y. Tang, R. Qiu, J. A. Fang, Q. Miao, and M. Xia, "Adaptive lag synchronization in unknown stochastic chaotic neural networks with discrete and distributed time-varying delays," *Physics Letters A*, vol. 372, no. 24, pp. 4425–4433, 2008.
- [22] Y. Xia, Z. Yang, and M. Han, "Lag synchronization of unknown chaotic delayed yang-yang-type fuzzy neural networks with noise perturbation based on adaptive control and parameter identification," IEEE Transactions on Neural Networks, vol. 20, no. 7, pp. 1165–1180, 2009.
- [23] Z. X. Liu, S. L. Liu, S. M. Zhong, and M. Ye, "p-th moment exponential synchronization analysis for a class of stochastic neural networks with mixed delays," *Communications in Nonlinear Science and Numerical Simulation*, vol. 15, pp. 1899–1909, 2010.
- [24] O. M. Kwon and J. H. Park, "Delay-dependent stability for uncertain cellular neural networks with discrete and distribute time-varying delays," *Journal of the Franklin Institute*, vol. 345, no. 7, pp. 766– 778, 2008.
- [25] R. Samidurai, S. Marshal Anthoni, and K. Balachandran, "Global exponential stability of neutral-type impulsive neural networks with discrete and distributed delays," *Nonlinear Analysis: Hybrid Systems*, vol. 4, no. 1, pp. 103–112, 2010.
- [26] Y. Horikawa and H. Kitajima, "Bifurcation and stabilization of oscillations in ring neural networks with inertia," *Physica D*, vol. 238, no. 23-24, pp. 2409–2418, 2009.
- [27] H. Lu, "Chaotic attractors in delayed neural networks," *Physics Letters A*, vol. 298, no. 2-3, pp. 109–116, 2002.
- [28] W. H. Chen and X. Lu, "Mean square exponential stability of uncertain stochastic delayed neural networks," *Physics Letters A*, vol. 372, no. 7, pp. 1061–1069, 2008.
- [29] H. Huang and G. Feng, "Delay-dependent stability for uncertain stochastic neural networks with time-varying delay," *Physica A*, vol. 381, no. 1-2, pp. 93–103, 2007.
- [30] T. Li, A. Song, S. Fei, and T. Wang, "Global synchronization in arrays of coupled Lurie systems with both time-delay and hybrid coupling," *Communications in Nonlinear Science and Numerical Simulation*, vol. 16, no. 1, pp. 10–20, 2011.
- [31] D. Yue, E. Tian, Y. Zhang, and C. Peng, "Delay-distribution-dependent stability and stabilization of T-S fuzzy systems with probabilistic interval delay," *IEEE Transactions on Systems, Man, and Cybernetics Part B*, vol. 39, no. 2, pp. 503–516, 2009.
- [32] X. Mao, Stochastic Differential Equations and Their Applications, Horwood, Chichester, UK, 1997.
- [33] Z. Wang, H. Shu, Y. Liu, D. W. C. Ho, and X. Liu, "Robust stability analysis of generalized neural networks with discrete and distributed time delays," *Chaos, Solitons & Fractals*, vol. 30, no. 4, pp. 886– 896, 2006.



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